

Stuff we missed --

5A-1

Gilbert cell (4D-e)

Math -- errors in the BFO (4B-3,4)

New: "Quadrature AM"

Hw:
4.8-1,2
4.4-1
4.3-7

Quadrature AM --

What it does:

Allows 2 signals to share a channel, allowing twice as many channels to fit in the same spectrum.

How?

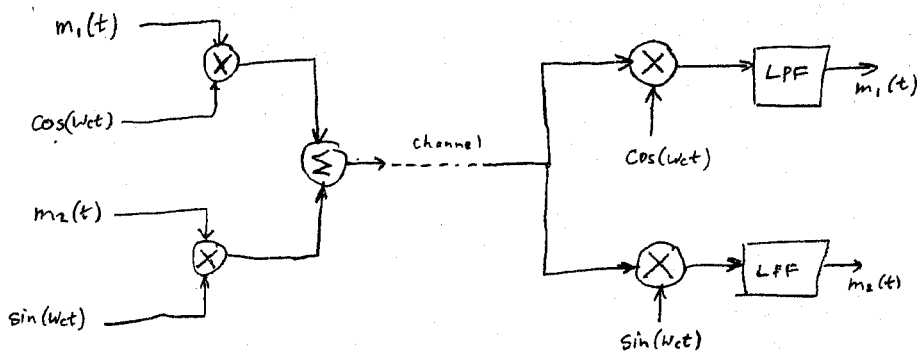
Use 2 carriers at the same frequency, 90° apart.

To transmit --
Combine them, and feed both to the power amplifier and antenna.

To receive --

Use 2 detectors, with their BFO's 90° apart. One is in phase-lock with each transmitted signal.

$$\phi_{\text{QAM}}(t) = \underbrace{m_1(t) \cos(\omega_c t)}_{\substack{\text{one signal} \\ \text{"I"} \\ \text{(in-phase)}}} + \underbrace{m_2(t) \sin(\omega_c t)}_{\substack{\text{the other signal} \\ \text{"Q"} \\ \text{(quadrature)}}$$



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Analysis - see 4B-4
Book does it in time domain.

Analysis - when it doesn't work well...

$\theta = \text{phase}$
 $= 0$ for I
 $\frac{\pi}{2}$ for Q

From book --

$$X_1(t) = [m_1(t) \cos(\omega_c t) + m_2(t) \sin(\omega_c t)] (2 \cos(\omega_c t + \theta))$$

by trig identity $\cos^2(x) = \frac{1}{2}(1 + \cos(2x))$

$$= m_1(t) \cos \theta + m_1(t) \cos(2\omega_c t + \theta) \leftarrow \text{desired signal}$$

$$- m_2(t) \sin \theta + m_2(t) \sin(2\omega_c t + \theta) \leftarrow \text{error "crosstalk"}$$

base band harmonic - filtered out.

cherry

After LPF:

$$m_1(t) = m_1(t) \cos \theta - m_2(t) \sin \theta$$

θ is the carrier phase error.

$$\text{crosstalk} = \frac{\sin \theta}{\cos \theta} = \tan \theta$$

This error is manageable with phase locked loop.

θ	$\tan \theta$	$20 \log_{10}(\tan \theta)$
1°	.0175	-35
2°	.0349	-29
5°	.0875	-21
10°	.176	-15

Bigger problem with QAM ---

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Selective fading ---

non-symmetry in pre-detection filter

or anything that changes

one sideband relative to the other

all cause crosstalk.

Analysis is difficult,

but consider a similar system

Suppose we want to send $m_1(t)$ and $m_2(t)$
using sum and difference..

$$s(t) = m_1(t) + m_2(t)$$

$$d(t) = m_1(t) - m_2(t)$$

on receive ---

hopefully $m_1(t) = \frac{s(t) + d(t)}{2}$

$$m_2(t) = \frac{s(t) - d(t)}{2}$$

Suppose $d(t)$ has a gain error of -0.01 (-0.08 dB) --
(Gain = 1.01 or .99)

To measure crosstalk, set $m_1(t) = 0$, $m_2(t) = 1$

so $s(t) = 1$, $d(t) = -1$

$$m_1(t) = \frac{s(t) + .99 d(t)}{2} = \frac{1 - .99}{2} = .01 \leftarrow \text{crosstalk}$$

$$20 \log(.01) = -40 \text{ dB}$$

- Corresponds to setting gain with 1% resistors.

Suppose error is -0.1 (-9 dB) (Gain = 0.9)

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$$m_1(t) = \frac{s(t) + .9 d(t)}{2} = \frac{0.1}{2} = .05$$

(corresponds to setting gain with 10% resistors)

$$20 \log(.05) = -26 \text{ dB}$$

This error can occur easily in the channel ---

We can manage the circuit by using precision components and good design.

We cannot improve the channel.

Another problem ---

Both signals must be generated together ---
in quadrature.

OK when signals are related (AM stereo, color TV)

but useless when signals are completely separate,
as in two different radio transmitters.

but both have this error so it's -20 dB.